# **EVALPFC2-ICE2PCS03**

300W PFC Evaluation Board with CCM PFC controller ICE2PCS03

Power Management & Supply



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300W PFC Evaluation Board with CCM PFC controller ICE2PCS03 License to Infineon Technologies Asia Pacific Pte Ltd

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#### 1 Content

The evaluation board presented here is a 300W power factor correction (PFC) circuit with 85~265VAC universal input and 393VDC fixed output. The continuous conduction mode (CCM) PFC controller ICE2PCS03 is employed in this board to achieve the unity power factor. This ICE2PCS03 is a design variant of ICE2PCS01 to incorporate the new input brown-out protection function. Appreciated for its high integrated design, ICE2PCS03 can achieve full requirements of the PFC application implemented in the 8-pin DIP8 and SO8 packages. At the same time the number of peripheral components is minimized. The operation frequency is fixed at 65kHz due to internal oscillator of ICE2PCS03. In order to improve the power conversion efficiency, the CoolMOS<sup>TM</sup> C3 series and high voltage silicon carbide (SiC) schottky diode thinQ!<sup>TM</sup> are used into this boost type PFC circuit.

#### 2 Evaluation Board





## 3 Technical Specifications

Input voltage	85VAC~265VAC
Input frequency	50Hz
Output voltage and current	393VDC, 0.75A
Output power	~ 300W
Efficiency	>91% at full load
Switching Frequency	100kHz

## 4 Circuit Description

#### Line Input

The AC line input side comprises the input fuse F1 as over-current protection. The high frequency current ripple is filtered by R1, L1 and CX1. The choke L2, X2-capacitors CX1 and CX2 and Y1-capacitor CY1 and CY2 are used as radio interference suppressors. RT1 is placed in series to limit inrush current during each power on.

#### Power Stage – Boost Type PFC Converter

After the bridge rectifier BR1, there is a boost type PFC converter consisting of L3, Q1, D1 and C2. The third generation CoolMOS™ SPP20N60C3 is used as the power switch Q1. BR1, Q1 and SiC Diode D1 share the same heat sink so that the system heat can be equably spread. Output capacitor C2 provides energy buffering to reduce the output voltage ripple (100Hz) to the acceptable level.

#### PWM Control of Boost Converter

The PWM control is realized by 8-Pin CCM PFC IC ICE2PCS03. It is a variant design of ICE2PCS01 with preserving most of the features. Unlike the conventional PFC controller, ICE2PCS03 does not need direct sine wave reference signal. The switching frequency is fixed at 65kHz by the IC internal oscillator. There are two control loops in the circuit, voltage loop and current loop. The output voltage is sensed by the voltage divider of R5A, R5B, R6A and R6B and sent to internal error amplifier. The output of error amplifier is used to control current in the inner current loop. The compensation network C4, C5, R7 constitutes the external circuitry of the error amplifier. This circuitry allows the feedback to be matched to various load conditions, thereby providing stable control. In order not to make the response for 100Hz ripple, the voltage loop compensation is implemented with low bandwidth. The inner loop, current control loop, is implemented with average current mode strategy. The instant current is adjusted to be proportional to both of MOSFET off duty DOFF and the error amplifier output voltage of voltage loop. The current is sensed by shunt resistors R2, R2A and R2B and fed into IC through R9. The current sense signal is averaged by an internal operating amplifier and then processed in the PWM generator which drives the gate drive. The averaging is realized by charging and discharging an external capacitor C7 at pin ICOMP.

The IC supply is provided by external voltage source and filtered and buffered by C8 and C9. The IC output gate driver is a fast totem pole gate drive. It has a built-in cross conduction current protection and a Zener diode to protect the external transistor switch against undesirable over voltages. The gate drive resistor R4 is selected to limit and gate pulse current and drive MOSFET for fast switching.

## **5** Circuit Operation

#### Soft Startup



When Vcc pin is higher than turn-on threshold, typical 11V, PFC is going to start. The unique soft start is integrated. Input current keeps sinusoidal and is increasing gradually until output voltage reaches 80% of rating. The boost diode is not stressed with large diode duty cycle under high current.

#### **Enhanced Dynamic Response**

Due to inherent low bandwidth of PFC dynamic, in case of load jump, regulation circuit can not response fast enough and it will lead to large output voltage overshoot or drop. To solve this problem in PFC application, enhanced dynamic response is implemented in the IC. Whenever output voltage exceeds by  $\pm 5\%$ , it will bypass the slow compensation operating amplifier and act on the nonlinear gain block to affect the duty cycle directly. The output voltage can be recovered in a short time.

#### **Protection Features**

#### a. Input brown-out protection

The dedicated input voltage brown-out VINS pin is the most distinct new feature brought by ICE2PCS03. This VINS pin senses a filtered input voltage divider and detects for the input voltage brown-out condition. If the detected VINS is below 0.8V, then IC output will be shut down. Only when VINS voltage reaches 1.5V can awake the IC again. Be informed that it will still have the soft start property when the IC is recovered from brown-out situation.

#### b. Open loop protection

The open loop protection is available for this IC to safe-guard the output. Whenever VSENSE voltage falls below 0.6V, or equivalently VOUT falls below 20% of its rated value, it indicates an open loop condition (i.e. VSENSE pin not connected). In this case, most of the blocks within the IC will be shutdown. It is implemented using a comparator with a threshold of 0.6V.

#### c. Output over-voltage protection

Whenever VOUT exceeds the rated value by 8%, the over-voltage protection OVP is active. This is implemented by sensing the voltage at pin VSENSE with respect to a reference voltage of 3.25V. A VSENSE voltage higher than 3.25V will immediately block the gate signal.

#### d. Soft over current control (SOC) and peak current limit

When the amplitude of current sense voltage reaches 0.68V, Soft Over Current Control (SOC) is activated. This is a soft control does not directly switch off the gate drive but acts on the internal blocks to result in a reduced PWM duty cycle.

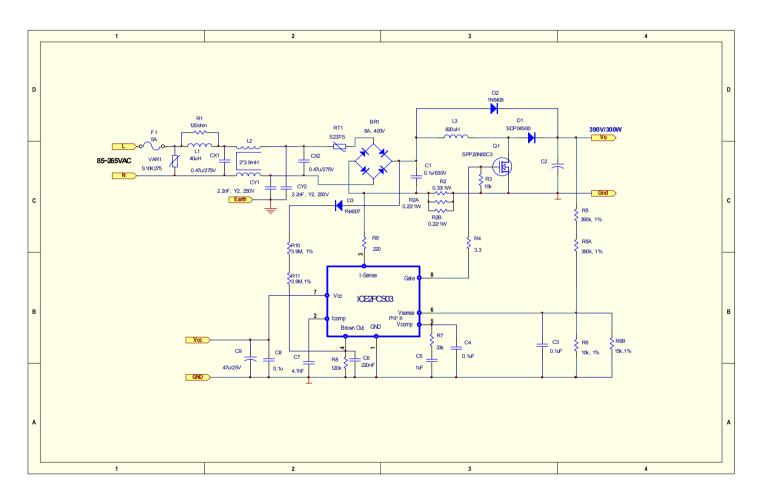
The IC also provides a cycle by cycle peak current limitation (PCL). It is active when the voltage at current sense voltage reaches -1.04V. The gate output is immediately off after 300ns blanking time.

#### e. IC supply under voltage lockout

When VCC voltage is below the under voltage lockout threshold VCCUVLO, typical 11, IC is off the gate drive is internally pull low to maintain the off state. The current consumption is down to 200uA only.



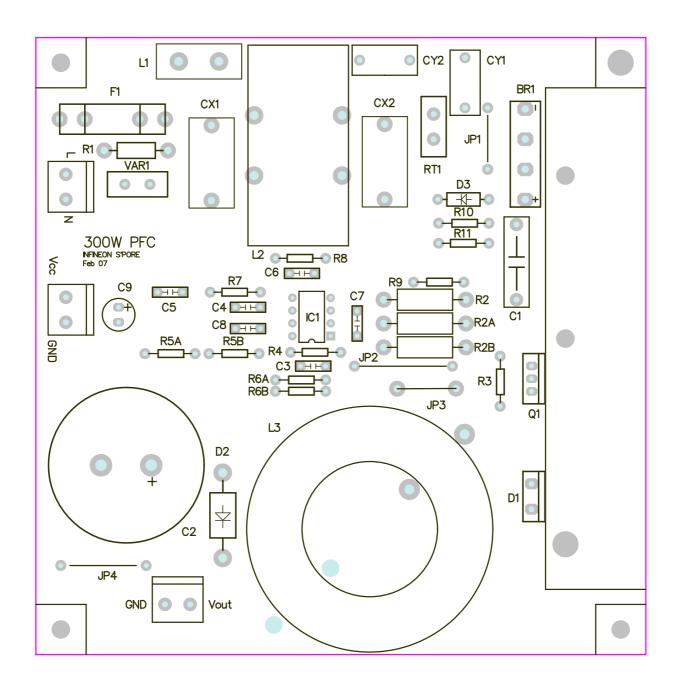
# 6 Circuit Diagram



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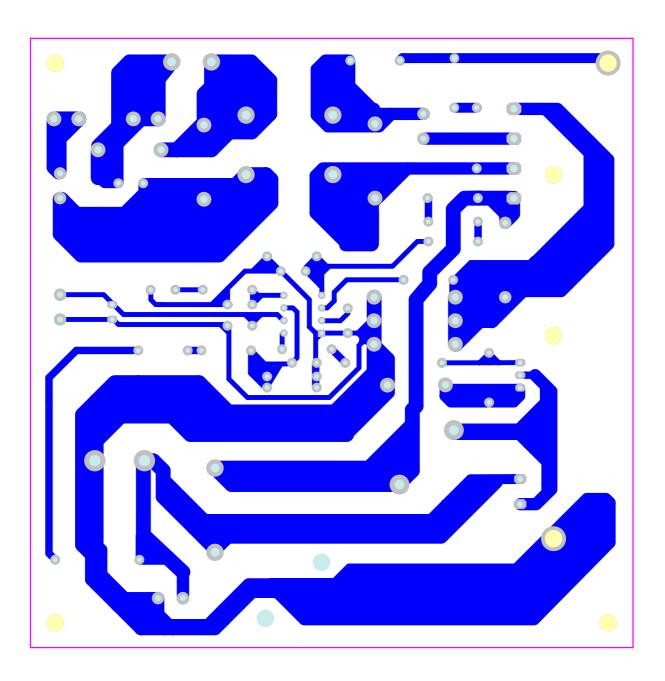


# 7 PCB Layout Top Layer





# 8 PCB layout Bottom Layer





# 9 Component List

Designator	Part Type	Description	Quantity	
BR1	8A, 400V	Bridge Rectifier	1	
C1	0.1uF/630V Ceramic Cap		1	
C2	220uF/450V	Electrolytic Cap	1	
C3*	0.1uF/50V	Ceramic Cap	1	
C4	0.1uF/50V	Ceramic Cap	1	
C5	1uF/50V	Ceramic Cap	1	
C6	220nF/50V	Ceramic Cap	1	
C7	4.7nF/50V	Ceramic Cap	1	
C8	0.1uF/50V	Ceramic Cap	1	
C9	47uF/25V	Electrolytic Cap	1	
CX1	0.47uF, X1, 305V	Ceramic Cap	1	
CX2	0.47uF, X1, 305V	Ceramic Cap	1	
CY1	2.2nF, Y2, 250V	Ceramic Cap	1	
CY2	2.2nF, Y2, 250V	Ceramic Cap	1	
		Connector	3	
D1	SDT04S60	Diode	1	
D2	1N5408	Diode	1	
D3	1N4007	Diode	1	
F1	5A	Fuse	1	
<del>-</del> 		Fuse Holder	2	
IC1	ICE2PCS03		1	
JP1	12.5mm, Ф0.7mm	Jumper	1	
JP2	20mm, Φ0.7mm	Jumper	1	
JP3	12mm, Φ1.2mm	Jumper	1	
JP4	17.5mm, Ф0.7mm	Jumper	1	
L1*	Shorted	· ·	0	
L2	2*3.9mH	CM Choke	1	
L3	820uH	Choke	1	
Q1	SPP20N60C3	Power MOSFET	1	
		Heat Sink	1	
		TO220 Clip	2	
		TO247 Clip	1	
		TO220 Isolation Pad	2	
		3mm Screw	3	
R1*	Not Connected		0	
R2	0.33/1W, 5%	Metal Film Resistor	1	
R2A	0.22/1W, 5%	Metal Film Resistor	1	
R2B	0.22/1W, 5%	Metal Film Resistor	1	
R3	10k/0.25W, 5%	Carbon Film Resistor	1	
R4	3.3/0.25W, 5%	Carbon Film Resistor	1	
R5A	390k/0.25W, 1%	Carbon Film Resistor 1		
R5B	390k/0.25W, 1%	Carbon Film Resistor 1		
R6A	10k/0.25W, 1%			
R6B	15k/0.25W, 1%	Carbon Film Resistor	1	
R7	33k/0.25W, 5%	Carbon Film Resistor	1	
R8	120k/0.25W, 1%	Carbon Film Resistor	1	



R9	220/0.25W, 5%	Carbon Film Resistor	1
R10	3.9M/0.25W, 1%	Carbon Film Resistor	1
R11	3.9M/0.25W, 1%	Carbon Film Resistor	1
RT1	S237/5	NTC Thermistor	1
VAR1	S10K275	Varistor	1

# 10 Boost Choke Layout

Core: CS400125 toriod

Turns: 70

Wire: 1 x Φ1.0mm, AWG19 Inductance: L=820uH

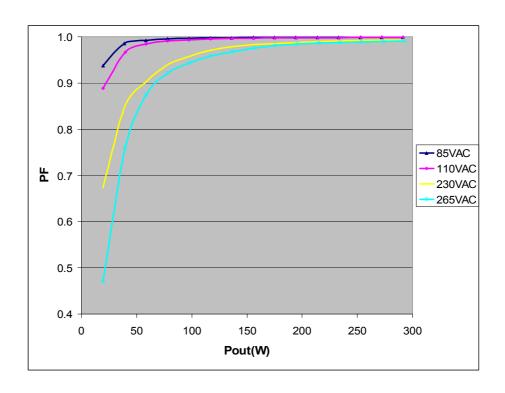
# 11 Test report

# 11.1 Load test (table and figure)

Vin(VAC)	Pin	lin	Vout	lout	Pout	Efficiency	PF
	318	3.756	388.3	0.75	291	91.58	0.9986
	252.8	2.972	388.6	0.6	233	92.23	0.9987
	208	2.449	388.9	0.5	194	93.49	0.9988
	168.2	1.98	389.2	0.4	156	92.56	0.9984
	146.5	1.727	389.4	0.35	136	93.03	0.998
	126.8	1.504	389.5	0.3	117	92.15	0.9976
	105.2	1.247	389.8	0.25	97	92.63	0.9968
	85.67	1.015	390	0.2	78	91.05	0.9953
	64.05	0.7578	390.3	0.15	59	91.41	0.9924
	44.47	0.533	390.7	0.1	39	87.86	0.9855
85VAC	22.3	0.278	391.4	0.05	20	87.76	0.938
	310.4	2.84	389	0.75	292	93.99	0.9985
	246.5	2.251	389.3	0.6	234	94.76	0.9983
	205.1	1.871	389.5	0.5	195	94.95	0.9977
	166.2	1.514	389.8	0.4	156	93.81	0.9969
	144.8	1.324	389.9	0.35	136	94.24	0.9962
	123.3	1.128	390.1	0.3	117	94.91	0.995
	103.9	0.951	390.3	0.25	98	93.91	0.9934
	84.58	0.776	390.4	0.2	78	92.31	0.9906
	63	0.581	390.7	0.15	59	93.02	0.984
	43.2	0.406	391	0.1012	40	91.6	0.967
110VAC	21.46	0.22	391.7	0.05	20	91.26	0.8892
230VAC	302.4	1.328	390.6	0.75	293	96.88	0.9937
	240.9	1.061	390.8	0.6	234	97.33	0.991
	203.1	0.896	390.9	0.5	195	96.23	0.9883
	162.8	0.723	391.1	0.4	156	96.09	0.9835
	142.1	0.631	391.2	0.35	137	96.35	0.9789



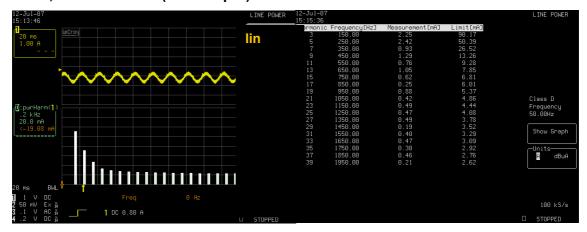
	122.7	0.551	391.3	0.3	117	95.67	0.9711
	101.7	0.463	391.4	0.25	98	96.21	0.9575
	82.6	0.383	391.5	0.2	78	94.79	0.94
	61.18	0.295	391.7	0.15	59	96.04	0.9034
	42.15	0.216	391.9	0.1012	40	94.09	0.8522
	21.1	0.136	392.1	0.05	20	92.91	0.674
	301.3	1.15	390.8	0.75	293	97.28	0.9913
	242.1	0.927	391	0.6	235	96.9	0.9879
	199.8	0.768	391.1	0.5	196	97.87	0.984
	160	0.62	391.3	0.4	157	97.83	0.975
	140.9	0.551	391.4	0.35	137	97.22	0.9678
	122.3	0.482	391.5	0.3	117	96.03	0.9588
	101.1	0.405	391.6	0.25	98	96.83	0.9436
	82.4	0.338	391.7	0.2	78	95.07	0.9209
	60.9	0.263	391.8	0.15	59	96.5	0.8758
	42	0.208	392	0.1012	40	94.45	0.7612
265VAC	21	0.17	392.2	0.05	20	93.38	0.47



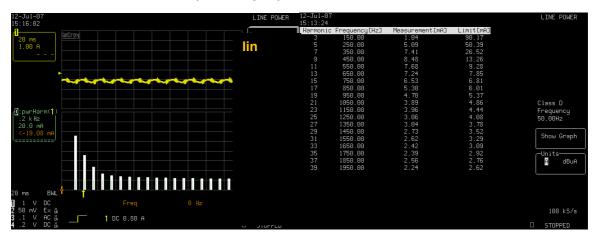


### 11.2 Harmonic test according to EN61000-3-2 Class D requirement

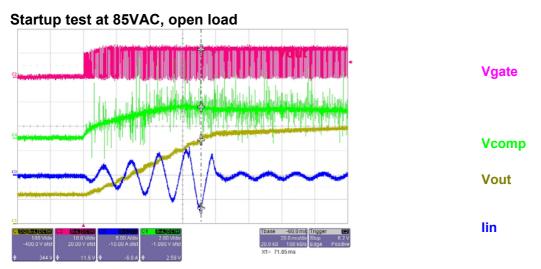
#### 85VAC, 7% of full load (22W output)



## 265VAC, 2% of full load (7W output)

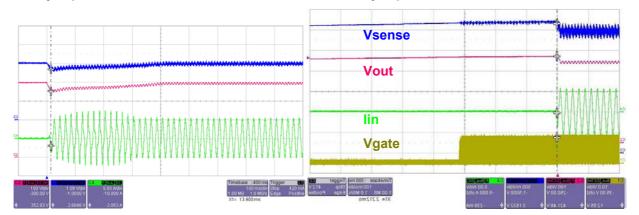


#### **Test Waveforms**



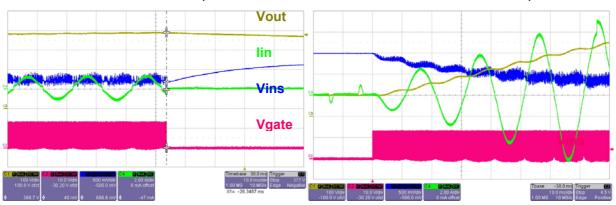


#### Load jump test at 85VAC, lout from 0A to 0.75A Load jump test at 85VAC, lout from 0.75A to 0A

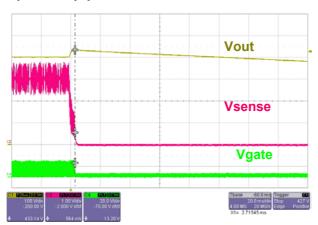


#### Enter brown-out at lout=0.1A, 56VAC

#### Leave brown-out at lout=0.1A, 74VAC



## Open Loop protection at 265V, lout=0.1A



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